

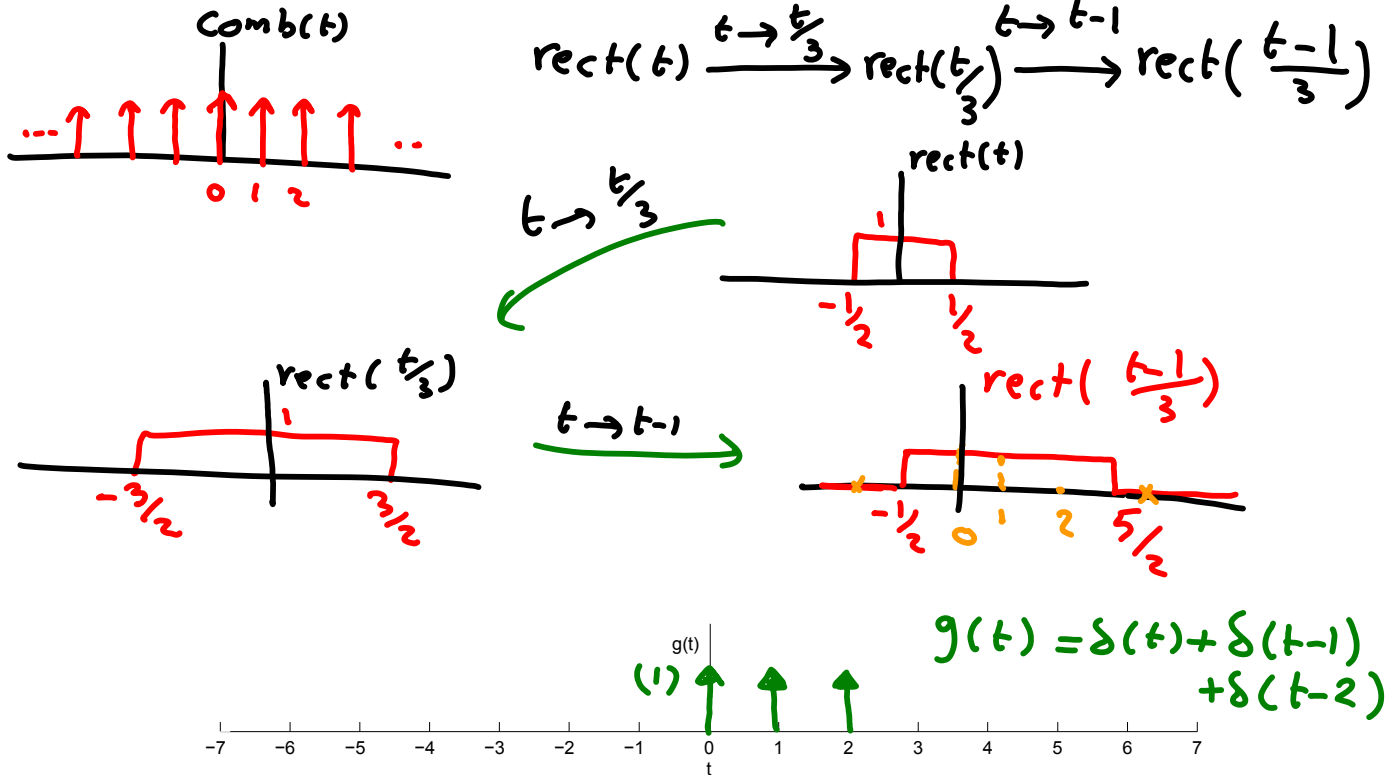
ENSC-380, Spring 2009  
Review Problems

1. (Midterm, 2008)

- a) Sketch the following signal:

$$g(t) = \text{comb}(t) \cdot \text{rect}\left(\frac{t-1}{3}\right)$$

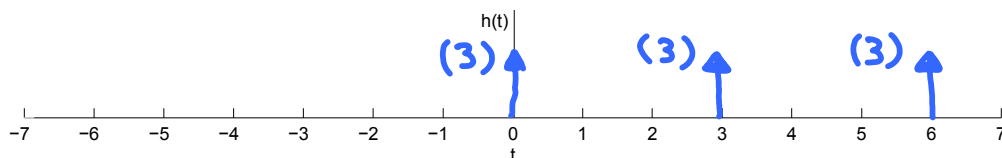
Use the space below for your intermediate work but show your final answer on the provided axes. Clearly indicate the values on both axes.



- b) Now sketch  $h(t) = g(t/3)$

$$g(t) = \delta(t) + \delta(t-1) + \delta(t-2) \Rightarrow h(t) = g\left(\frac{t}{3}\right) = \delta\left(\frac{t}{3}\right) + \delta\left(\frac{t}{3}-1\right) + \delta\left(\frac{t}{3}-2\right)$$

$$\Rightarrow h(t) = \delta\left(\frac{t}{3}\right) + \delta\left(\frac{t-3}{3}\right) + \delta\left(\frac{t-6}{3}\right) = 3\delta(t) + 3\delta(t-3) + 3\delta(t-6)$$



2. (Midterm 2008) A CT-LTI system is defined with the following differential equation:

$$y'(t) + 2y(t) = x(t) + 2x'(t)$$

Find the impulse response of this system,  $h(t)$ .  $h'(t) + 2h(t) = \delta(t) + 2\delta'(t)$

1)  $t < 0 \Rightarrow h(t) = 0$

2)  $t > 0 \Rightarrow h'(t) + 2h(t) = 0 \Rightarrow h(t) = k_1 e^{\alpha t}$

$$\cancel{k_1 \alpha e^{\alpha t}} + 2 \cancel{k_1 e^{\alpha t}} = 0 \Rightarrow \alpha = -2 \Rightarrow h(t) = k_1 e^{-2t}$$

3) Is  $h(t) = k_1 e^{-2t} u(t)$  enough? No

$$\Rightarrow h(t) = k_1 e^{-2t} u(t) + k_2 \delta(t)$$

replace in diff. equ. for  $t=0$ :

$$-2k_1 e^{-2t} u(t) + k_1 e^{-2t} \delta(t) + k_2 \delta'(t)$$

$$+ 2k_1 e^{-2t} u(t) + 2k_2 \delta(t) = \delta(t) + 2\delta'(t)$$

at  $t=0$

$$\begin{cases} k_1 + 2k_2 = 1 \\ k_2 = 2 \end{cases} \Rightarrow \begin{cases} k_1 = -3 \\ k_2 = 2 \end{cases} \Rightarrow h(t) = -3 e^{-2t} u(t) + 2\delta(t)$$

3. (Midterm 2008) Use the properties of CTFS and the FS formula given below ()

$$\underline{\cos(2\pi f_0 t)} \xleftrightarrow{FS} \frac{1}{2} \{ \delta[k-1] + \delta[k+1] \} \quad \text{Representation Period: } T_0 = \frac{1}{f_0}$$

to find the CTFS representation (Harmonic function) of

$$x(t) = \cos(200\pi t + \pi/4)$$

with representation period of  $T_f = 1/50$ .

**Notes:** 1) Do not use any other entries from the FS table. 2) Try to simplify your result as much as possible.

$$x(t) = \cos(2\pi \times 100 \times t + \frac{\pi}{4}) = \cos(2\pi \times 100 (t + \frac{1}{800}))$$

$$\cos(2\pi \times 100 \times t) \xleftrightarrow{FS} \frac{1}{2} \{ \delta[k-1] + \delta[k+1] \} \quad Y[k]$$

$T_f = T_0 = \frac{1}{100}$   
 or  
 $f_f = 100$

Change of Rep. Period:  $\cos(2\pi \times 100 \times t) \xleftrightarrow{FS} \begin{cases} Y[\frac{k}{2}] & k/2 \text{ integ.} \\ 0 & \text{else} \end{cases}$

$T_f = 2T_0 = \frac{1}{50}$

$$\cos(2\pi \times 100 \times t) \xleftrightarrow{FS} \frac{1}{2} \{ \delta[k-2] + \delta[k+2] \}$$

$T_f = \frac{1}{50}$

Time shift:  $\cos(2\pi \times 100 (t + \frac{1}{800})) \xleftrightarrow{FS}$

$T_f = \frac{1}{50}$   
 $f_f = 50$

$$X[k] = e^{+j2\pi k \times 50 \times \frac{1}{800}} \times \frac{1}{2} \{ \delta[k-2] + \delta[k+2] \}$$

$$X[k] = e^{j \frac{2\pi \times 2 \times 50}{800}} \times \frac{1}{2} \delta[k-2] \\ + e^{-j \frac{2\pi \times 2 \times 50}{800}} \times \frac{1}{2} \delta[k+2]$$

$$= \frac{1}{2} e^{j \frac{\pi}{4}} \delta[k-2] + \frac{1}{2} e^{-j \frac{\pi}{4}} \delta[k+2]$$

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$$\delta(at) = \frac{1}{a} \delta(t)$$

$$\text{but for DT: } \delta\left[\frac{n}{m}\right] = \delta[n]$$

4. Text, Problem 4-25 part (b)
5. Text, Problem 4-29
6. Prove the “Time Reversal” property of CTFS.

4.25 (b) Find the CTFS of

$$x(t) = 5[\text{tri}(t-1) - \text{tri}(t+1)] * \frac{1}{4} \text{comb}\left(\frac{t}{4}\right), T_F = 4$$

4.29 Find the signal power of  $x[n]$ :

$$x[n] = 5 \sin\left(\frac{14\pi n}{15}\right) - 8 \cos\left(\frac{26\pi n}{30}\right)$$

$$\left. \begin{aligned} x_1[n] &= \sin\left(\frac{14\pi n}{15}\right) \Rightarrow N_1 = 15 \\ x_2[n] &= \cos\left(\frac{2\pi \times 13 \times n}{30}\right) \Rightarrow N_2 = 30 \end{aligned} \right\} \Rightarrow N_f = 30$$

Find the smallest intg.  $N_1$ , s.t.

$$\sin\left(\frac{14\pi(n+N_1)}{15}\right) = \sin\left(\frac{14\pi n}{15}\right)$$

$$\frac{14\pi N_1}{15} = k \times 2\pi \Rightarrow 7N_1 = k \times 15$$

$$\begin{matrix} \downarrow & \downarrow \\ N_1 = 15 & 7 \end{matrix}$$

from Appendix E:

$$\cos\left(\frac{2\pi n}{N_0}\right) \xleftrightarrow{N_f = mN_0} \frac{1}{2} \left( \text{comb}_{N_f} [k-m] + \text{comb}_{N_f} [k+m] \right)$$

$$\sin\left(\frac{2\pi n}{N_0}\right) \xleftrightarrow{N_f = mN_0} \frac{j}{2} \left( \text{comb}_{N_f} [k+m] - \text{comb}_{N_f} [k-m] \right)$$

$N_0$  is not necessarily an integer!

$$x_1[n] = \sin\left(\frac{14\pi n}{15}\right) = \sin\left(\frac{2\pi n}{15/7}\right)$$

$$N_0 = \frac{15}{7}$$

$$\sin\left(\frac{2\pi n}{15/7}\right) \xleftrightarrow[N_f = 30 = 14N_0]{FS} \frac{j}{2} \left\{ \text{Comb}_{30}[k+14] - \text{Comb}_{30}[k-14] \right\}$$

$\uparrow$   
 $m$

$$x_2[n] = \cos\left(\frac{26\pi n}{30}\right) = \cos\left(\frac{2\pi n}{30/13}\right)$$

$\underbrace{\quad}_{N_0}$

$$x_2[n] \xleftrightarrow[N_f = 30 = 13N_0]{FS} \frac{1}{2} \left\{ \text{Comb}_{30}[k-13] + \text{Comb}_{30}[k+13] \right\}$$

$\uparrow$   
 $m$

⋮



Prove the 'time reversal' property for CTFS.

$$\textcircled{1} \quad x(t) = \sum_{k=-\infty}^{\infty} X[k] e^{j2\pi k f_0 t}$$

$$\text{suppose } y(t) = x(-t) = \sum_{k=-\infty}^{\infty} Y[k] e^{j2\pi k f_0 t} \quad \textcircled{2}$$

$$\textcircled{1} \Rightarrow x(-t) = \sum_k X[k] e^{-j2\pi k f_0 t}$$

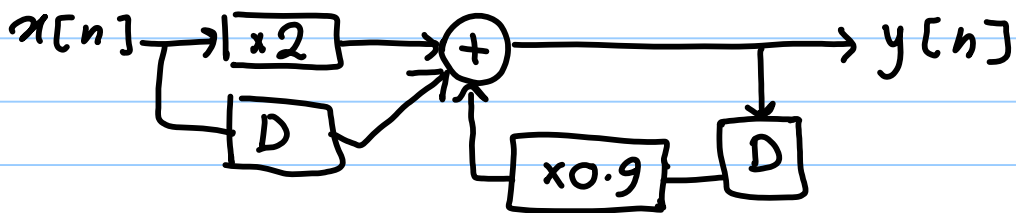
$$k \rightarrow -k \quad x(-t) = \sum_{k=-\infty}^{\infty} X[-k] e^{j2\pi k f_0 t} \quad \textcircled{3}$$

compare  $\textcircled{2}$  &  $\textcircled{3}$

$$\Rightarrow Y[k] = X[-k]$$

Thus if  $x(t) \xrightarrow{FS} X[k]$   
then  $x(-t) \xrightarrow{FS} X[-k]$

3.43 Find the impulse response of this system:



$$y[n] = 0.9 y[n-1] + 2x[n] + x[n-1]$$

$$\Rightarrow y[n] - 0.9 y[n-1] = 2x[n] + x[n-1] \quad (*)$$

Assume  $h_1[n]$  to be the impulse response to the following system:

$$y[n] - 0.9 y[n-1] = 2x[n]$$

Also assume that  $h_2[n]$  is the impl. resp. to another system:

$$y[n] - 0.9 y[n-1] = x[n-1]$$

Then the impulse response to the original system (\*) is  $h[n] = h_1[n] + h_2[n]$

on the other hand,  $h_2[n] = \frac{1}{2} h_1[n-1]$

So it is enough to find  $h_1[n]$ :

$$h_1[n] - 0.9 h_1[n-1] = 2\delta[n]$$

- For  $n < 0 \Rightarrow h_1[n] = 0$

- For  $n > 0 \Rightarrow h_1[n] = k (0.9)^n$

- at  $n = 0 \Rightarrow h_1[0] - 0.9 h_1[-1] = 2$

$\Rightarrow h_1[0] = 2 \Rightarrow k = 2 \Rightarrow h_1[n] = 2 (0.9)^n u[n]$

Now,  $h_2[n] = \frac{1}{2} h_1[n-1]$

$\Rightarrow h_2[n] = (0.9)^{n-1} u[n-1]$

Finally:  $h[n] = h_1[n] + h_2[n]$

$$= 2 \times (0.9)^n u[n] + (0.9)^{n-1} u[n-1]$$

$$= 2 \times (0.9)^0 \delta[n] + 2 \times (0.9)^n u[n-1] + (0.9)^{n-1} u[n-1]$$

$$= 2 \delta[n] + (0.9)^{n-1} u[n-1] (1.8 + 1)$$

$$= 2 \delta[n] + 2.8 \times (0.9)^{n-1} u[n-1]$$