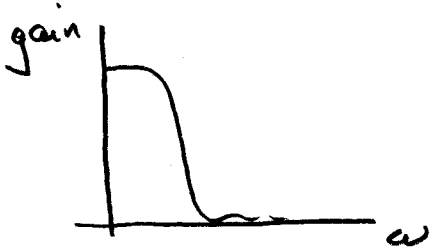


4 FREQUENCY RESPONSE

DC & L Chapter 15

- "Frequency response" refers to the steady-state response of the circuit to a sinusoidal input. It specifies the gain and phase shift relating output to input as a function of input frequency.
- Frequency response has this physical interpretation only for stable systems (poles in the LHP, or $\int_0^{\infty} |h(t)| dt < \infty$), for which the transients (contributions to the output from $H(s)$ poles) die out.
- The frequency response is as general a description of circuit behaviour as impulse response, step response, transfer function or DE, since any one can be obtained from any other.

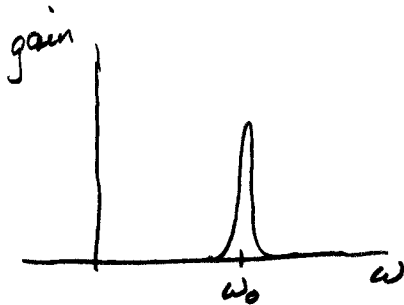
- We often characterise circuits by the gain portion of their frequency response:



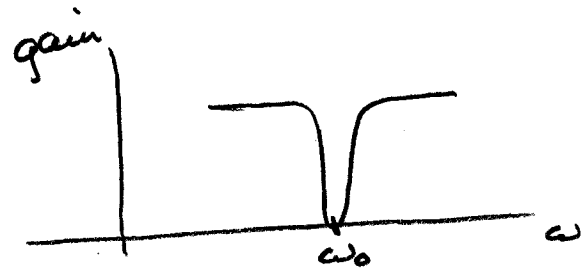
Lowpass: good for suppressing high freq. noise contaminating a low freq signal.



Highpass: good for removing unwanted DC offset; a change detector.



Narrow bandpass, tuned or resonant filter: good for extracting a periodic component in the input or for selecting one tone or channel.



Notch filter: good for removing unwanted tones or channels, or for detecting change from periodic input (with period $2\pi/\omega_0$)

but phase shift as a function of frequency is also important!

4.1 Frequency Response by Laplace Transform

- If $H(s)$ is the transfer function, $H(j\omega)$ is the frequency response: $|H(j\omega)|$ is the amplitude gain, $\angle H(j\omega)$ or $\arg[H(j\omega)]$ is the phase shift.

A demonstration of this below.

- Consider the input

$$x(t) = \cos(\omega t) u(t) \longleftrightarrow X(s) = \frac{s}{s^2 + \omega^2}$$

for some frequency ω . The output is

$$Y(s) = H(s) \frac{s}{s^2 + \omega^2}$$

$$y(t) = \sum_i \text{Res}_i(t) \quad \text{residues of } H(s) \frac{s e^{st}}{s^2 + \omega^2}$$

poles: $\pm j\omega$ plus those of $H(s)$

- By hypothesis, poles of $H(s)$ contribute transient terms, so steady state output

$$y(t) = 2 \operatorname{Re} \left[H(s) \frac{s e^{st}}{s^2 + \omega^2} \right]_{s=j\omega} = 2 \operatorname{Re} \left[H(j\omega) \frac{j\omega}{2j\omega} e^{j\omega t} \right]$$

$$= \operatorname{Re} \left[H(j\omega) e^{j\omega t} \right]$$

— As usual, we can write this in rectangular or polar.

$$y(t) = |H(j\omega)| \operatorname{Re}[e^{j\angle H(j\omega)} e^{j\omega t}]$$

$$= |H(j\omega)| \cos(\omega t + \angle H(j\omega)) \quad \text{usual}$$

or with $H(j\omega) = H_r(\omega) + j H_i(\omega)$,

$$y(t) = H_r(\omega) \cos(\omega t) - H_i(\omega) \sin(\omega t) \quad \text{less common}$$

• example



$$H(s) = \frac{1/sC}{R + 1/sC} = \frac{1/RC}{s + 1/RC} = \frac{1/\tau}{s + 1/\tau}$$

$$y(t) = \mathcal{L}^{-1}\left[H(s) \frac{s}{s^2 + \omega^2} \right] = \mathcal{L}^{-1}\left[\frac{s/\tau}{(s + 1/\tau)(s^2 + \omega^2)} \right]$$

$$= \operatorname{Re}[H(j\omega) e^{j\omega t}] - \underbrace{\frac{1}{1 + \tau^2 \omega^2} e^{-t/\tau}}_{\text{dies away}}$$

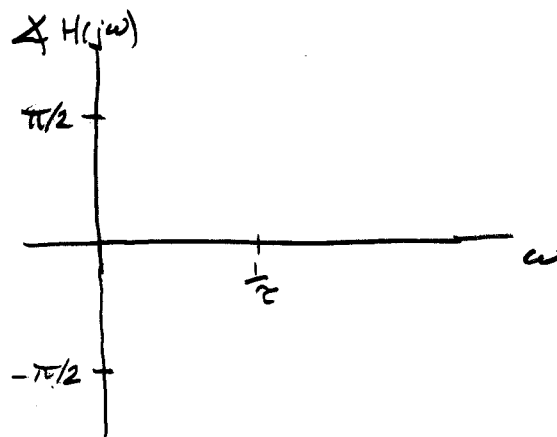
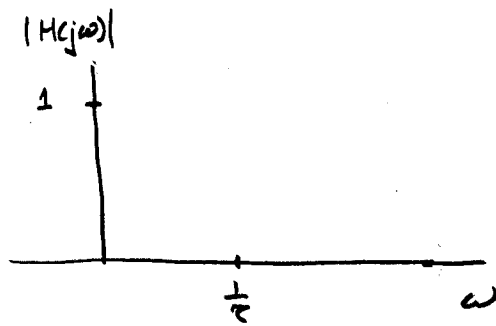
in steady state:

$$= \operatorname{Re}\left[\frac{1/\tau}{j\omega + 1/\tau} e^{j\omega t} \right]$$

$$= \frac{1}{\sqrt{1 + \omega^2 \tau^2}} \cos(\omega t - \tan^{-1}(\omega \tau))$$

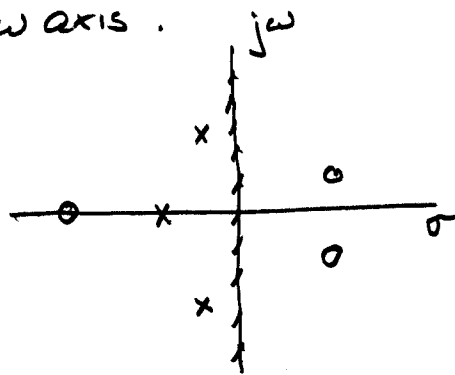
Its frequency response is $H(j\omega) = \frac{1/\tau}{j\omega + 1/\tau}$ or, equivalently,

the two real functions, like $|H(j\omega)|$ and $\angle H(j\omega)$



• Interpretation:

- Frequency response $H(j\omega)$ is $H(s)$ along the $j\omega$ axis.



The $j\omega$ axis is in the ROC, since all poles are in LHP

- Frequency response is Fourier transform of $h(t)$.

$$H(j\omega) = H(s)|_{s=j\omega} = \int_{0^-}^{\infty} h(t) e^{-j\omega t} dt$$

- Recover $H(s)$ from freq response with $\omega \leftarrow \frac{s}{j} = -js$

• More examples — elementary circuits.

— differentiator

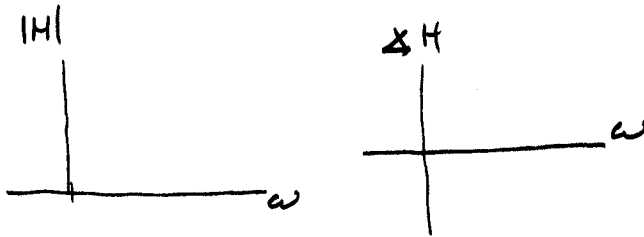
$$y(t) = \frac{d}{dt} x(t) \quad , \quad \text{for } x(t) = \cos(\omega t), \quad y(t) = -\omega \sin(\omega t)$$

$$Y(s) = s X(s) \quad \text{so } H(s) = s$$

Then frequency response is $H(j\omega) = j\omega$. We have

$$|H(j\omega)| = \omega \quad \angle H(j\omega) = \pi/2$$

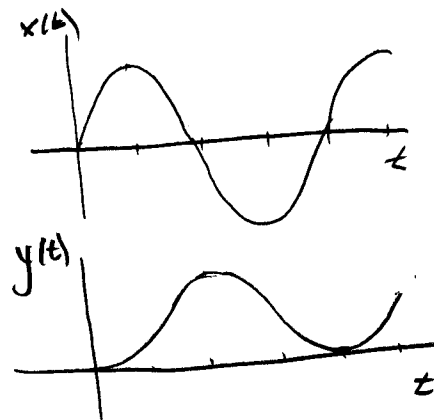
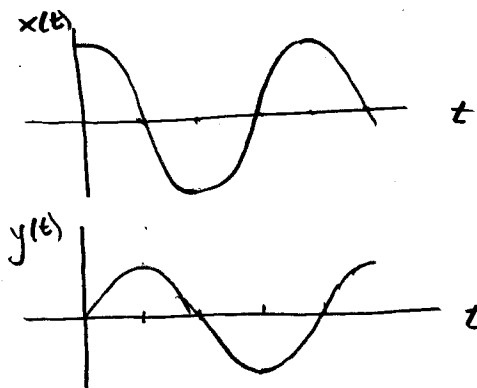
Do these values make sense?



— integrator, $y(t) = \int_0^t x(\alpha) d\alpha$; if $x(t) = \cos(\omega t)$, $y(t) = \frac{\sin(\omega t)}{\omega}$

$$Y(s) = \frac{1}{s} X(s) \quad \text{so } H(s) = \frac{1}{s}$$

Hmm. This pole is not in the LHP.

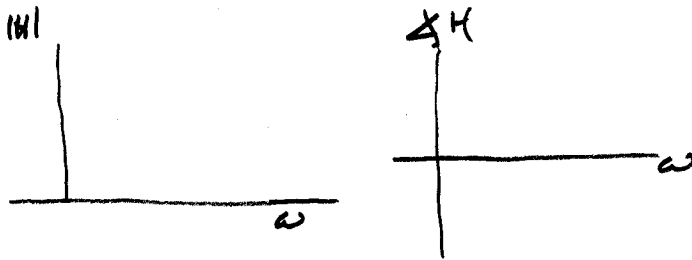


a persistent constant offset from pole at $s=0$

Sometimes we're willing to overlook the non-transient "transient" and loosely refer to the particular solution's frequency response

$$H(j\omega) = \frac{1}{j\omega}$$

Does it make sense?



— Delay device.

$$y(t) = x(t-T)$$

$$x(t) = \cos(\omega t),$$

$$y(t) = \cos(\omega(t-T))$$

$$= \cos(\omega t - \omega T)$$

We have $H(s) = e^{-sT}$ (not realizable by finite order)

so freq response $H(j\omega) = e^{-j\omega T}$

sensible?

