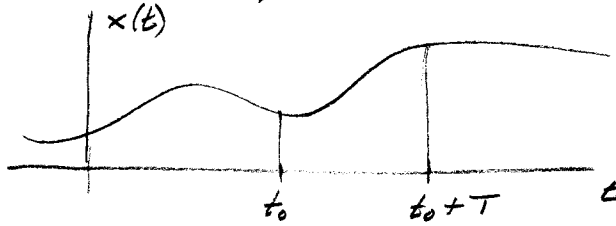


4.3 Fourier Series

H&V 3.3

read "complex exponentials and orthogonality" on website 4.3.1

Over a finite interval, a function can be expressed as sum of sinusoids, each with its own amplitude and phase, at frequencies that are multiples of a fundamental.



for $t_0 \leq t \leq t_0 + T$:

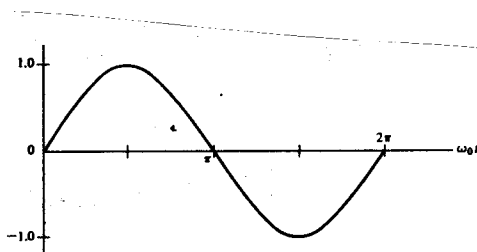
$$\hat{x}(t) = \sum_{n=0}^{\infty} R_n \cos\left(\frac{2\pi n t}{T} + \phi_n\right)$$

$$= \sum_{n=0}^{\infty} A_n \cos\left(\frac{2\pi n t}{T}\right) - B_n \sin\left(\frac{2\pi n t}{T}\right)$$

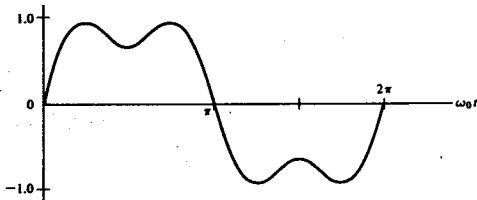
$$= \sum_{n=-\infty}^{\infty} X_n e^{j 2\pi n t / T}$$

$$(X_{-n} = X_n^* \text{ for real } x(t))$$

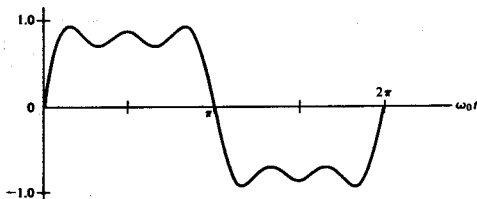
$$R_n e^{j\phi_n} = A_n + jB_n = X_n$$



(a) The first partial sum is a sine wave

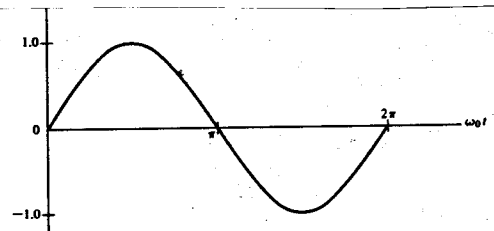


(b) The second partial sum begins to approximate a square wave

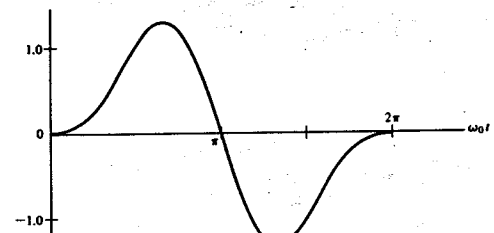


(c) The third partial sum provides a better approximation to a square wave

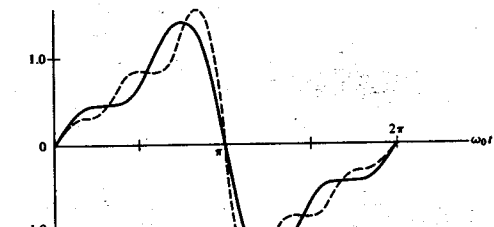
FIGURE 3-1. The series of Example 3-1 shows how a trigonometric series approximates a square wave (only one period shown).



(a) The first partial sum is a sine wave



(b) The second partial sum begins to approximate a saw tooth waveform



(c) The third and fifth partial sums better approximate a saw tooth waveform

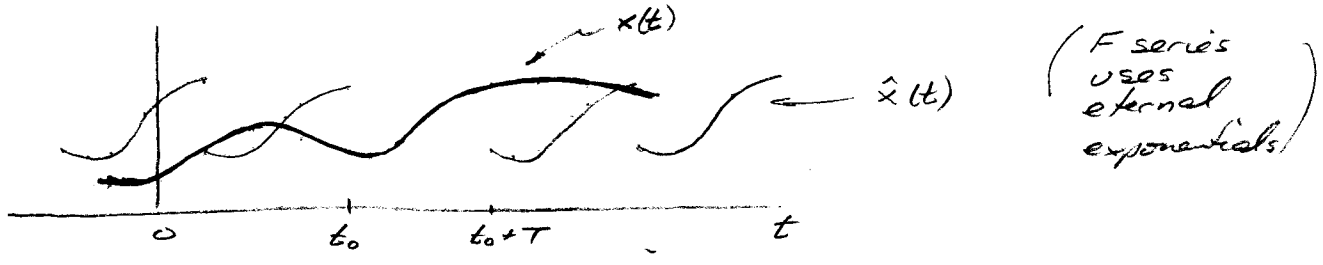
FIGURE 3-2. The trigonometric series of Example 3-2 illustrates approximation of a sawtooth waveform (only one period shown).

If we select the coefficients properly then:

the series $\hat{x}(t) = x(t)$ the original function, in $t_0 \leq t \leq t_0 + T$.

Outside the representation interval they may differ, since

$$\hat{x}(t) = \sum_{n=0}^{\infty} R_n \cos\left(\frac{2\pi n t}{T} + \phi_n\right) \text{ is periodic in } t, \text{ period } T$$



If $x(t)$ is itself periodic with period T , then $\hat{x}(t) = x(t)$.

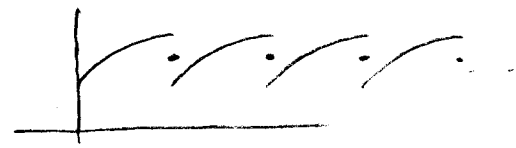
Actually $\hat{x}(t) \rightarrow x(t)$ in this sense:

$$\text{if } \hat{x}_n(t) = \sum_{k=0}^n R_k \cos(k\omega_0 t + \phi_k) \quad (\omega_0 = \frac{2\pi}{T}, f_0 = \frac{1}{T})$$

and the error $e_n(t) = x(t) - \hat{x}_n(t)$

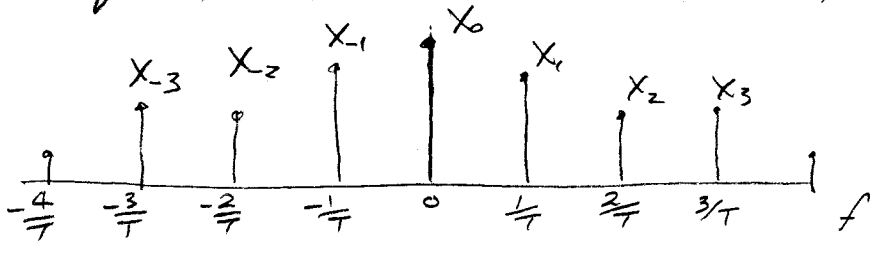
$$\text{then } \lim_{n \rightarrow \infty} \int_{t_0}^{t_0+T} e_n^2(t) dt = 0$$

At discontinuities, it converges to average of left and right limits:



Can have only a finite number of step discontinuities in the representation interval.

The frequency components are spaced by $\frac{1}{T} = f_0$ Hz
 $\frac{2\pi}{T} = \omega_0$ rps



- Here's how to calculate the coefficients, Easiest in exponential form:

$$x(t) = \sum_{k=-\infty}^{\infty} X_k e^{jk\omega_0 t} = \hat{x}(t)$$

$$x(t) e^{-jn\omega_0 t} = \sum_{k=-\infty}^{\infty} X_k e^{j(k-n)\omega_0 t}$$

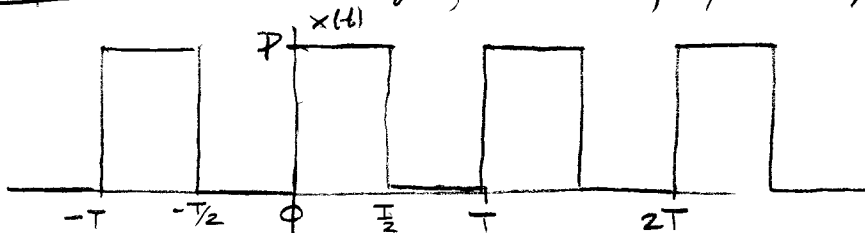
$$\int_{t_0}^{t_0+T} x(t) e^{-jn\omega_0 t} dt = \sum_{k=-\infty}^{\infty} X_k \underbrace{\int_{t_0}^{t_0+T} e^{j(k-n)\omega_0 t} dt}_{\begin{matrix} T & k=n \\ 0 & k \neq n \end{matrix}}$$

$$X_n = \frac{1}{T} \int_{t_0}^{t_0+T} x(t) e^{-jn\omega_0 t} dt = T X_n$$

$$x(t) = \sum_{n=-\infty}^{\infty} X_n e^{jn\omega_0 t}$$

Fourier series pair

example one/zero square wave, peak to peak P



choose symmetric interval $(-\frac{T}{2}, \frac{T}{2})$ for convenience

$$X_n = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jn\omega_0 t} dt = \frac{P}{T} \int_0^{T/2} e^{-jn2\pi t/T} dt$$

$$= \frac{P/T}{-jn2\pi/T} (e^{-jn\pi} - 1)$$

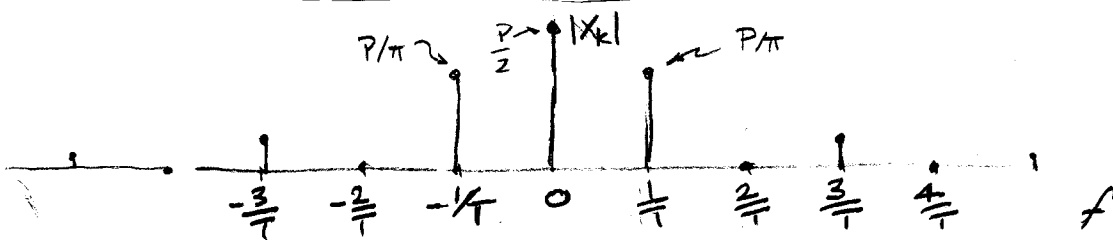
$$= \begin{cases} P/2 & n=0 \\ 0 & \text{other even } n \\ -\frac{jP}{\pi n} & n \text{ odd} \end{cases}$$

$$\text{so } x(t) = \frac{P}{2} + \sum_{\text{odd } n} \left(\frac{-jP}{\pi n} \right) e^{j2\pi n t / T}$$

$$= \frac{P}{2} + \sum_{\text{odd } n > 0} 2 \operatorname{Re} \left[\frac{-jP}{\pi n} e^{j2\pi n t / T} \right]$$

since the $-n$ term is the conjugate of the $+n$ term

$$= \frac{P}{2} + \sum_{\text{odd } n > 0} \frac{2P}{\pi n} \sin\left(2\pi n \frac{t}{T}\right)$$



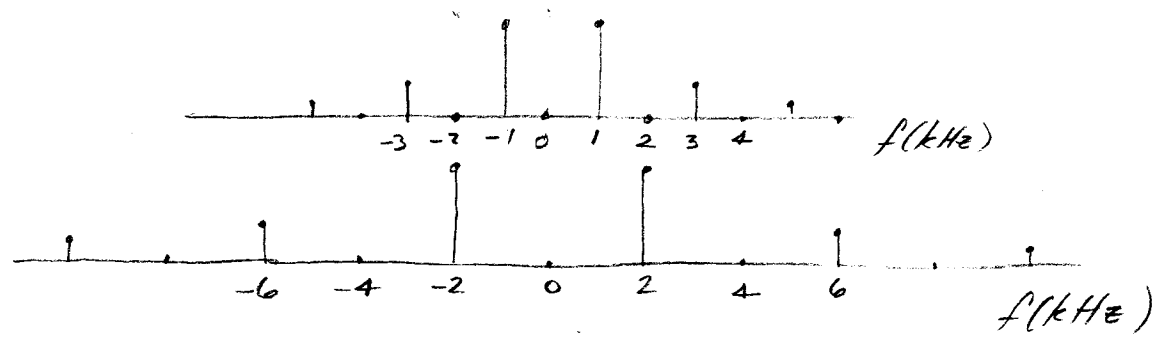
- The sketch shows the amplitude distribution in f
 - no even harmonics for symmetric square wave
 - odd harmonics are in proportion $1, \frac{1}{3}, \frac{1}{5}, \frac{1}{7}, \dots$
 - the component at $f=0$ is the dc, or average, value

- In terms of amplitude $A = \frac{1}{2} P$ (the pk-pk) we

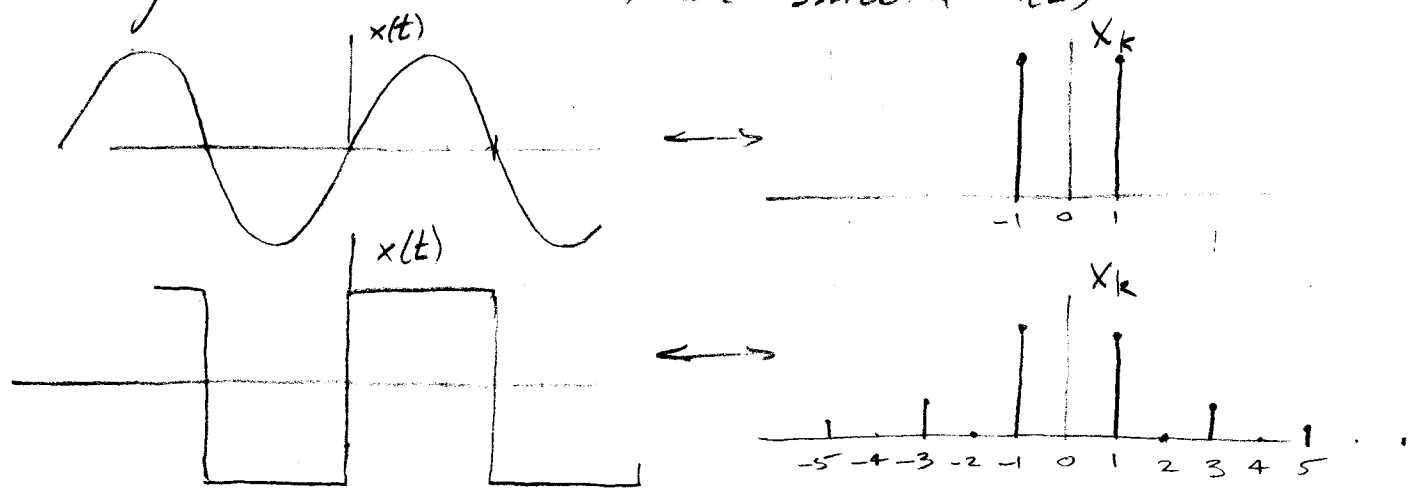
have

$$x(t) = A + \sum_{\text{odd } n > 0} \frac{4A}{\pi n} \sin\left(2\pi n \frac{t}{T}\right)$$

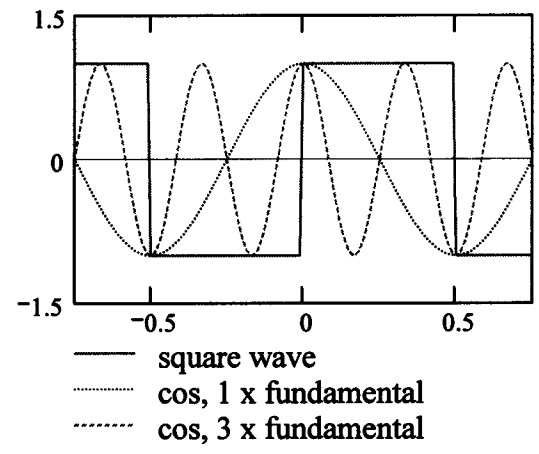
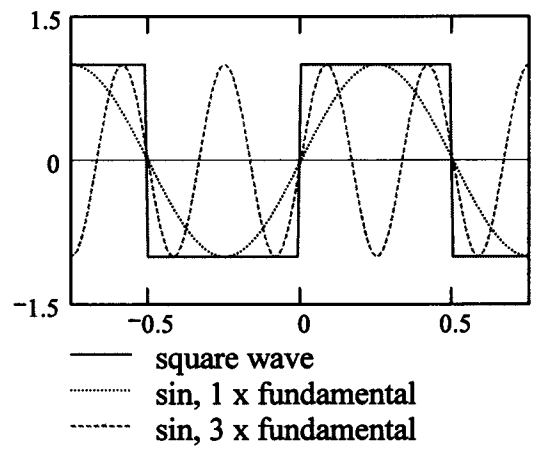
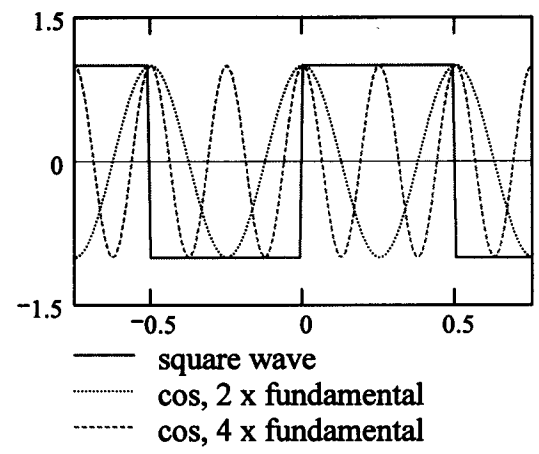
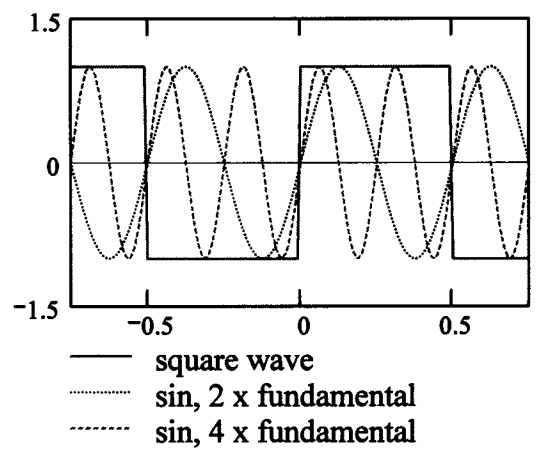
- Quickly varying $x(t)$ have greater bandwidth than slowly varying:
 - A 1kHz periodic signal has components at multiples of 1kHz. Speed it up to 2kHz and frequency components spread out twice as far



- Even with the same period, jagged $x(t)$ have greater bandwidth than smooth $x(t)$

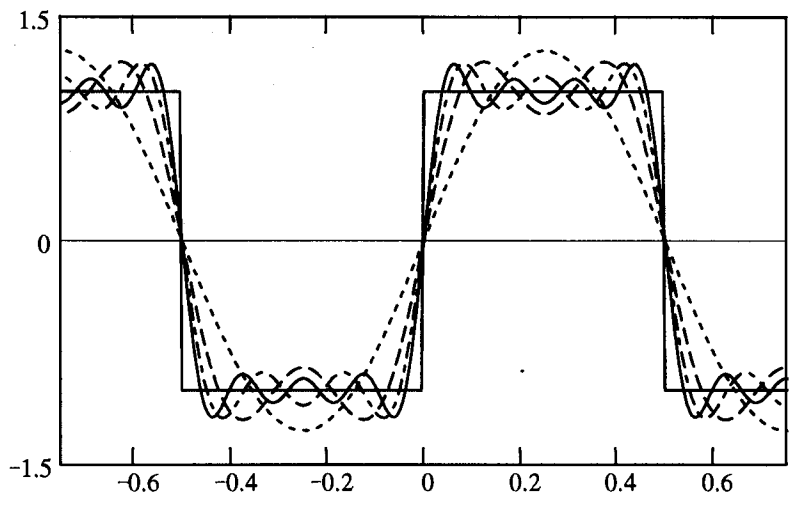


Only the odd coefficients of a square wave are non-zero:

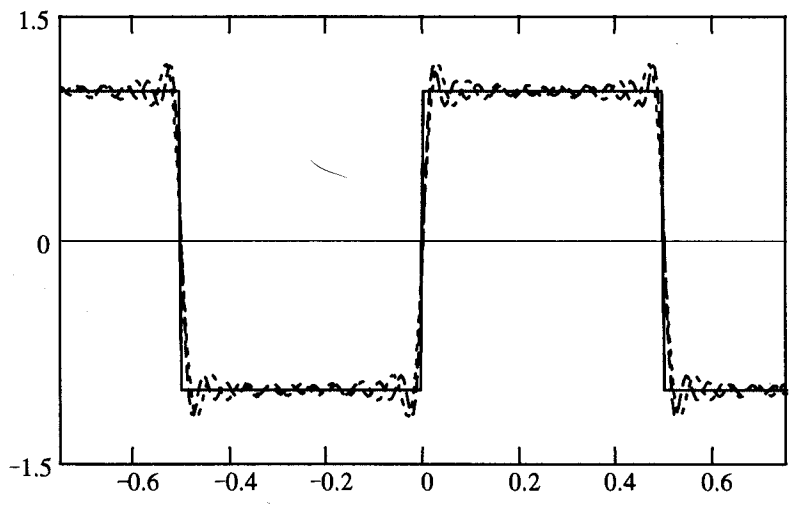


An illustration of the Gibbs phenomenon:

$$P = 1 \quad \hat{x}(t, N) := \sum_{n=1}^N \text{odd}(n) \cdot \frac{4 \cdot P}{\pi \cdot n} \cdot \sin\left(2 \cdot \pi \cdot n \cdot \frac{t}{T}\right)$$



- square wave
- - - n=1 term
- - - terms 1 and 3
- - - terms 1,3 and 5
- terms 1,3,5 and 7



- square wave
- - - ~~n=1 term~~ 15
- - - ~~terms 1 and 3~~ 21