ENSC327 Communications Systems 25: ISI and Pulse Shaping (Ch 6)

Jie Liang School of Engineering Science Simon Fraser University

Outline

□ Chapter 6: Baseband Data Transmission

- No modulation, sending pulse sequences directly
- Suitable for lowpass channels (eg, coaxial cables)
- (Chap 7 will study digital bandpass modulation, for bandpass channels (eg, wireless) that need high-freq carrier)

□ ISI and Pulse Shaping

- ISI Definition
- Zero ISI Condition
- Nyquist Pulse Shaping Condition
- Nyquist bandwidth
- Nyquist channel

Introduction

- A digital communication system involves the following operations:
 - Transmitter: maps the digital information to analog electromagnetic energy.
 - Receiver: records the analog electromagnetic energy, and recover the digital information.
- □ The system can introduces two kinds of distortions:
 - Channel noise: due to random and unpredictable physical phenomena.
 Studied in Chap. 9 and 10.
 - Intersymbol interference (ISI): due to imperfections in the frequency response of the channel.
 - The main issues is that previously transmitted symbols affect the current received symbol.
 - **Studied in Chap. 6.**



- **Input**: binary data, bk = 0 or 1, with duration T_b.
 - Bit rate: 1/Tb bits/second.
- □ Line encoder (Chap. 5): electrical representation of the binary sequences, e.g.,

$$a_k = \begin{cases} +1, & b_k = 1, \\ -1, & b_k = 0. \end{cases}$$



- □ Transmit filter: use pulses of different amplitudes to represent one or more binary bits.
- \square The basic shape is represented by a filter g(t) or G(f)
- □ The output discrete PAM signal from the transmitter:



□ Channel:

- If the channel is ideal, no distortion will be introduced.
- A practical channel can be represented by a linear, time invariant (LTI) filter.
- The channel output: the convolution of the transmitted signal and the channel impulse response. This introduces intersymbol interference (ISI)



- **Receiver** can be represented by another filter:
 - To remove noise and cancel channel distortion.
- **The output is**
- The output is then sampled synchronously with the clock at the transmitter.
- Finally, a decision-making device is used to recover the binary bits. Different methods can be used, e.g.,
 - Threshold, equalization, maximal likelihood decoding

6.1 Baseband Transmission



- The model can be used to study baseband system as well as bandpass system (via complex envelope)
- Objectives of pulse shaping:

 \Box Given h(t), design g(t) and q(t) to eliminate the intersymbol interference (ISI).

□The transmitted signal should have small bandwidth to meet the bandwidth constraint of the channel.

Equivalent Model

The joint effect of g(t), h(t), and q(t) is a composite filter:

→ Equivalent system model:



□ What is the condition on p(t) to eliminate ISI?

Mathematical Definition of ISI

source
$$\sum_{k} a_k \delta(t - kT_b)$$
 \Longrightarrow $y(t) = \sum_{k} a_k p(t - kT_b)$

Zero ISI Condition

■ To get $\sum_{k \neq i} a_k p_{i-k} = 0$ for any ak, the sample of p(t) should satisfy $p_i = p(iT_b) = \begin{cases} \sqrt{E}, & i = 0, \\ 0, & i \neq 0. \end{cases}$

□ This is true if, for example,

$$p(t) = \sqrt{E}\operatorname{sinc}(\frac{t}{T_b}).$$



Zero ISI Condition

Example: A series of sinc pulses corresponding to the sequence 1011010. No ISI at sampling times.



Zero ISI Condition



□ Is sinc function a good composite filter? No!

- Infinite duration
- Non-causal
- A small timing jitter can cause large ISI.
- \Box Any other possibilities for p(t)?
 - Yes. There are infinite number of solutions!
 - What is the general condition for p(t) to be ISI free?

The condition
$$p_i = p(iT_b) = \begin{cases} \sqrt{E}, & i = 0, \\ 0, & i \neq 0. \end{cases}$$
 is achieved

if the frequency response of p(t) satisfies

$$\sum_{k=-\infty}^{\infty} P(f + \frac{k}{T_b}) = T_b \sqrt{E}.$$

(The sum of all shifted versions of P(f) by multiple of

1 / Tb should be constant at all frequencies).



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Proof:

Note that $\sum_{k=-\infty}^{\infty} P(f + \frac{k}{T_b})$ is periodic signal with period 1 / Tb



The condition is not difficult to meet (e.g., by raised cosine pulse in Chap 6.4)

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Nyquist Channel and Nyquist Bandwidth

- Special case:
- $B_0 = \frac{1}{2T_b}$ If the bandwidth of P(f) is then to satisfy

$$\sum_{k=-\infty}^{\infty} P(f + \frac{k}{T_b}) = T_b \sqrt{E},$$

P(f) must be

■ If the bandwidth of P(f) is less than 1 / (2Tb),

■ If the bandwidth of P(f) is greater than 1 / (2T_b),

$$\sum_{k=-\infty}^{\infty} P(f + \frac{k}{T_b}) = T_b \sqrt{E}$$

Nyquist channel and Nyquist bandwidth:

□ If the bit duration is fixed at T_b, or bit rate R_b is 1/T_b, then the transmission bandwidth is

$$B_T \ge \frac{1}{2T_b} = \frac{1}{2} R_b$$

so the minimal bandwidth is half of the bit rate.

On the other hand, if the transmission bandwidth is fixed at BT, then

$$R_b \leq 2B_T$$

the maximal transmitted bits per second (bit rate) is twice of the bandwidth.

f

t=Tb=1/(2B₀)

 $T_b \sqrt{E}$

Bo

-B0

p(t)