# ENSC327 Communication Systems26: Raised Cosine Pulse and Eye Diagram

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### Outline

**□** 6.4 Raised cosine pulse spectrum

6.6 Eye Diagram

# Nyquist Pulse Shaping Condition

$$
\sum_{k=-\infty}^{\infty} P(f + \frac{k}{T_b}) = T_b \sqrt{E}.
$$

 $\Box$ Nyquist channel and Nyquist bandwidth:

> If the bit duration is fixed at  $T_b$ , or bit rate Rb is 1/Tb, then the transmission bandwidth is

$$
B_T \ge B_0 = \frac{1}{2T_b} = \frac{1}{2} R_b
$$

so the minimal bandwidth is half of the bit rate.

П On the other hand, if the transmission bandwidth is fixed at BT, then

$$
R_b \leq 2B_T
$$

 $\Box$  the maximal transmitted bits per second (bit rate) is twice of the bandwidth.

t

f

 $T_b \sqrt{E}$ 

 $B_0=1/(2T_b)$ 

)  $t=Tb=1/(2B_0)$ 

-B0

 $p(t)$ 

## Examples of zero ISI spectrum

If BW of P(f) is greater than  $1/(2Tb)$ , there are infinite possible solutions to satisfy

$$
\sum_{k=-\infty}^{\infty} P(f + \frac{k}{T_b}) = T_b \sqrt{E}.
$$

In particular, if the BW of  $P(f)$  is less than  $1/Tb$ , then the summation only involves two terms at each frequency, which can be easily satisfied:



# Raised Cosine Spectrum

 $\Box$  For example, the previous case can be achieved by using raised cosine function.



Many other functions also satisfy the requirement, for example,

P(f) can be a triangle. But the cosine function has some nice properties.

#### Raised Cosine Spectrum

**□** The previous example can be generalized. **□ Raised Cosine Spectrum:** Let  $2B_0 = 1/T_b$ , or  $T_b \sqrt{E} = \frac{\sqrt{E}}{2B_0}$  $T_b\sqrt{E}=\frac{V}{2}$ E

> : cut  $f_1$  : cut-off frequency.  $f_1$ : cut-off frequency.<br> $\begin{array}{c} \textbf{Math functions:} \\ \textbf{r}\textbf{cosfir} \textbf{if} \textbf{$

rcosine( ): FIR/IIR

E

=

#### Example: Raised Cosine Spectrum



7( $\frac{1}{2(B_0 - f_1)}$ : )- $\binom{0-f_1}{0}$  $\frac{|J|}{B_0 - f_1}$  $\pi$  $\int$  $\frac{f_1}{f_1}$ : scales [f<sub>1</sub>, 2B<sub>0</sub> – f<sub>1</sub>] to [0,  $\pi$ ].  $\Box$  Constant for  $f < f1$ , raised cosine for f in [f<sub>1</sub>, 2B<sub>0</sub> – f<sub>1</sub>]. ( $\frac{1}{2(B_0 - f_1)}$ : )-) $1 + \cos$  $0$   $J_1$  $\overline{1}$  $\int$  $\bigg)$  $\setminus$  $\bigg($ −−  $+$  cos  $2(B_0 - f_1)$ π $f$ |- $f_1$  $\frac{N_{1}}{N_{2}}$ : Mapped to  $1 + \cos(x)$ , x in [0,  $\pi$ ].

## Example: Raised Cosine Spectrum

Roll-off factor, or excess-bandwidth factor (over Nyquist bw B0):

Cut-off frequency:



Bandwidth:

- $\Box$ f1 and  $\alpha$  can be adjusted to control the trade-off between the bandwidth and length of the impulse response.
- $\alpha > 0$ : more bandwidth than Nyquist, but filter is shorter.

#### Example: Raised Cosine Spectrum  $f_1 = (1 - \alpha)B_0$  $-\alpha$ ) $B$

 A nice thing about raised cosine window is that its impulse responsehas closed-form expression:

This is a scaled sinc function:  $a = 0: \rightarrow p(t) = \sqrt{E}$ sinc $(2B_0 t)$ .



- If rolloff factor  $\alpha$  increases
	- $\blacksquare$  f<sub>1</sub> decreases
	- T Bandwidth increases
	- T But  $p(t)$  is shorter  $\rightarrow$  More robust to timing error. robust to timing error.

#### Root Raised Cosine Pulse

 $\Box$  If the raised cosine pulse is used, we have  $p(t)$  $= g(t) * h(t) * q(t)$ =Raised cosine function $\Box$  One way to achieve this is:

- Given the channel  $H(f)$ , use the first equation to find transmitter filter G(f).
- Received filter is the root raised cosine.

#### Example: Bandwidth of T1 system

- **□** T1 system: multiplexing 24 telephone inputs.
- **Bit duration:**  $T_b = 0.647 \,\mu s$  $=0.647\,\mu$
- $\Box \rightarrow$  Bit rate: = $=1/$ = $R_{b} = 1/T_{b} = 1.544 \text{ Mb/s}$
- The Nyquist bandwidth is (minimal required BW):

 $\Box A$  more realistic choice is to use  $\alpha = 1$ :

 $\Box$ Note: the unit of bandwidth is Hz, and the unit of bit rate is bits/sec (b/s).

### Outline

# **□** 6.4 Raised cosine pulse spectrum

6.6 Eye Diagram

# 6.6 Eye Diagram

### An effective way to observe ISI



- $\mathcal{C}_{\mathcal{A}}$ Extract one or more symbol periods
- $\mathcal{L}_{\mathcal{A}}$ Superimpose all possible results
- $\mathcal{L}_{\mathcal{A}}$ Can be easily obtained by oscilloscope





http://www.highfrequencyelectronics.com/Archives/Nov05/HFE1105\_Tutorial.pdf

#### cont …

# Easy to show on a scope



Figure  $1 \cdot$  At top is an undistorted eye diagram of a band-limited digital signal. The bottom eye pattern includes amplitude (noise) and phase (timing) errors. The various transition points can provide insight into the nature of the impairments.

# Example: sinc pulse

 $\Box$  If the composite filter  $p(t)$  is a sinc pulse:

 $\Box$  If only interferences from the immediate neighboring pulses are considered: ==> 8 possibilities



t

**Th** 

 $p(t)$ 

## Example: sinc pulse



17

# Example: sinc pulse

Details within [-Tb/2, Tb/2]



If there is no noise and no timing error, the data 1 or -1 can be perfected detected at time 0.

# Eye Diagram Summary

# **□** Practical eye diagrams have some errors:



FIGURE 6.6 Interpretation of the eye pattern for a baseband binary data transmission system.